



IP-16DAC

Three Channel 16-bit Digital to Analog
Converter for Instrumentation
IndustryPack™

User Manual

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Three Channel Digital to Analog
Converter for Instrumentation
IndustryPack™

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Product Description

IP-16DAC provides three independent channels of high resolution 16-bit digital to analog conversion. Applications include instrumentation, precision positioning, calibration, remote programming. Factory calibration and low drift produce high accuracy and high repeatability for precision laboratory applications. Four output ranges, including bipolar and unipolar, plus remote sense, provide direct interface to target hardware in most cases.

Sophisticated features make the IP-16DAC versatile and easy to interface in demanding jobs. The digital voltages (DAC registers) are readable by software. The DACs may be cleared by software or reset to either bipolar or unipolar zero. External power supplies may be used to minimize system noise. The DACs include internal de-glitching hardware that produces a faster settling time.

Factory Calibration traceable to the National Bureau of Instruments is standard. Zero offset is calibrated to ± 1 LSB. Gain is calibrated to $\pm 0.008\%$. Output is always monotonic. Calibration consists of correction parameters burned into a custom PROM on each IndustryPack. Driver software reads the calibration parameters to correct digital data prior to writing to the DAC.

Worst case differential non-linearity over temperature is ± 1 LSB. This exceptional linearity may be used to digitally construct low-distortion waveforms. DC drift with temperature is typically 2 ppm/ $^{\circ}$ C for both the DAC and the on board voltage reference. Since the average power dissipation for the fully loaded IndustryPack is less than one watt, normal temperature drift is negligible.

A very high precision on board reference is used to provide exceptional stability. Software autocalibration provides a continuous accuracy of plus or minus one LSB (least significant bit).

The IP-16DAC conforms to the IndustryPack Interface Specification. This guarantees compatibility with multiple carrier boards, busses and form factors. This compatibility permits application code migration between otherwise incompatible systems. This provides for a longer life of the customers' software investment.

Connection to the IP-Precision ADC is via a standard 50-conductor ribbon cable. Four parallel wires are used for each buffered output to reduce wiring resistance. Buffered virtual ground connections exist between channels to reduce channel to channel crosstalk. A low current sense input is provided to minimize total system error. This cable may be terminated in a 50 position screw terminal block, available from SBS Technologies, or any user-determined hardware. Provision is made to permit external power supply connections if desired.

Address Map for VMEbus

IP-16DAC is accessed using 16-bit words only. It may be accessed in either the I/O space or the memory space of the carrier board. The data written to each DAC may be read back for software convenience, debugging, or calibration.

Shown below in Figure 1 is the memory map of the IP-16DAC for the VMEbus. See the following sections for addressing in AT bus or NuBus systems.

VMEbus Address	Access Mode	Function
Base + 0	read/write	DAC A
Base + 2	read/write	DAC B
Base + 4	read/write	DAC C

Note: All accesses are word wide (16-bit).
Note: Base address is set on carrier board.

Figure 1 Address Map of IP-16DAC for VMEbus

The physical address of each DAC is the sum of the base address of the VMEbus carrier board, plus the offset of the IndustryPack slot, plus the offset of the individual DAC.

Note that the standard IP-16DAC is accessible in both the I/O and memory space. Each hardware write to the DAC initiates a sequence. The DAC is selected, its output is held using the internal track and hold circuit, data is written to the DAC input buffer, then the data is transferred the DAC itself. Finally the internal track and hold goes back to track mode.

After this sequence, the output will settle to its final value. See the specification section below for settling time information. Note also that even with the internal track and hold there is some output glitching with each transition, even minor ones.

Some operating systems, such as OS-9, search for memory at reset. The IP-16DAC *must not* appear in the memory search space (often A24, 00000 to 7FFFFFFF). The readback capability will confuse the memory search into thinking that the IP-16DAC is general purpose read/write memory. Placing the IP-16DAC in the I/O space and *disabling all memory accesses to the VMEbus IndustryPack carrier* will in most cases provide a safe and convenient location.

AT (ISA) bus Addressing

SBS' ATC4 AT bus carrier permits up to four IndustryPacks to be installed into one PC bus slot.

The ATC4 has a one-time set up program called ATCSETUP (previously called SPSETUP) that is provided on floppy disk with each ATC4. This program has a "fill in the blank" menu that the AT systems integrator uses to select the address location and mode for the ATC hardware. When the ATCSETUP program terminates it creates and installs a device driver called SOLUPACK.SYS in the CONFIG.SYS. This device driver programs the control registers on the ATC4 for proper operation each time the system is booted.

After the driver is run at boot time, the ATC4 board is normally "transparent" to the user's software. Application software directly addresses the I/O registers on each IndustryPack.

The base address of the four IPs on the ATC4 is specified in the ATCSETUP menu. Offsets from this base address are shown in hex in Figure 2. The additional offsets for accessing each DAC are shown below in Figure 3. Note that the physical address of each DAC is the sum of three numbers: the IP_Base_Address (from the ATCSETUP), the IP Offset (from Figure 2), and the relay offset (from Figure 3).

IP Module	IP I/O Offset Addresses
IP A	IP_Base_Address + 0
IP B	IP_Base_Address + \$80
IP C	IP_Base_Address + \$100
IP D	IP_Base_Address + \$180

Note: IP_Base_Address is set in the ATSETUP menu.

Figure 2 AT bus Base Addressing

IP-16DAC is accessed using 16-bit words only. It may be accessed in either the "IOSel" or "MEMSel" space of the ATC. (Note that this is different than programming the ATC4 to be in the I/O space or memory space of the AT bus.) The IOSel space is recommended as being standard. The data written to each DAC may be read back for software convenience, debugging, or calibration. The ATC4 *must* be programmed for 16-bit operation.

Shown below in Figure 3 is the map of the IP-16DAC for the AT bus.

VMEbus Address	Access Mode	Function
IP Offset + 0	read/write	DAC A
IP Offset + 2	read/write	DAC B
IP Offset + 4	read/write	DAC C

Note: All accesses are word wide (16-bit).
Note: IP Offset address is from Figure 2.

Figure 3 Address Map of IP-16DAC for AT bus

Note that the standard IP-16DAC is accessible in both the IOSel and the MemSel space on the ATC4. IOSel access is standard.

Each hardware write to the DAC initiates a sequence. The DAC is selected, its output is held using the internal track and hold circuit, data is written to the DAC input buffer, then the data is transferred the DAC itself. Finally the internal track and hold goes back to track mode.

After this sequence, the output will settle to its final value. See the Specifications Section below for settling time information. Note also that even with the internal track and hold there is some output glitching with each transition, even minor ones.

Software wishing to access the ID PROM must first access the ATC4 control register, switching the mode from I/O read/write to ID read. (D4,D3 of Carrier_Base_Address changes from 00 to 01). See the ATC4 User Manual for more information. Software which accesses the ID PROM should restore the ATC4 control register when it completes.

Address Map for NuBus

IP-16DAC is accessed using 16-bit words only. It may be accessed in either the I/O space or the memory space of the carrier board. The data written to each DAC may be read back for software convenience, debugging, or calibration.

Shown below in Figures 4 and 5 are the memory maps of the IP-16DAC for the RM1260 and RM1270. See the previous section for addressing in VMEbus systems.

RM1260 Address	Access Mode	Function
base + 0	read/write	DAC A
base + 4	read/write	DAC B
base + 8	read/write	DAC C

Note: All accesses are word wide (16-bit).

Figure 4 Address Map of IP-16DAC for NuBus RM1260

RM1270 Address	Access Mode	Function
base + 2	read/write	DAC A
base + 6	read/write	DAC B
base + A	read/write	DAC C

Note: All accesses are word wide (16-bit).

Figure 5 Address Map of IP-16DAC for NuBus RM1270

Each hardware write to the DAC initiates a sequence. The DAC is selected, its output is held using the internal track and hold circuit, data is written to the DAC input buffer, then the data is transferred the DAC itself. Finally, the internal track and hold goes back to track mode.

After this sequence, the output will settle to its final value. See the specification section below for settling time information. Note also that even with the internal track and hold there is some output glitching with each transition, even minor ones.

Programming

Programming the DAC typically involves a simple 16-bit write. Sample drivers are available for several host systems. Sample code for the 68020 is available in the form of an Applications Note. Note that drivers should use the calibration information contained in the on board PROM to meet specified accuracy. Information on how to use the calibration parameters is contained in the Calibration section, below.

Because of the internal track and hold amplifier, each DAC has an effective maximum update rate of 500 KHz. Tight loops that attempt to write to a DAC faster than this will execute, but the DAC output will stay in the hold mode until the loop terminates. Only the last value to be written to the DAC will make it to the analog output.

The three DACs are independent. That is, writing to one has no effect (including no output glitch) on the other two.

The physical address of each DAC is the sum of the base address of the carrier board, plus the offset of the IndustryPack slot, plus the offset of the individual DAC.

Some operating systems, such as OS-9, search for memory at reset. In these cases the IP-16DAC *must not* appear in the memory search space. The readback capability will confuse the memory search into thinking that the IP-16DAC is general purpose read/write memory.

Contact SBS Technologies for more information.

ID PROM

Every IP contains an IP PROM, whose size is at least 32 x 8 bits. The ID PROM aids in software auto configuration and configuration management. The user's software, or a supplied driver, may verify that the device it expects is actually installed at the location it expects, and is nominally functional. The ID PROM contains the manufacturing revision level of the IP. If a driver requires that a particular revision be present, it may check for it directly.

Standard Data for the ID PROM on the IP-16DAC is shown in Figure 6, below. For more information on IP ID PROMs refer to the IndustryPack Interface Specification.

(The addresses in the Figure are for VMEbus systems. For raw PROMs—in a programmer, for example—subtract one and divide the address by two. For NuBus systems, subtract one and multiply by two, then add one for RM1260 or three for the RM1270.)

The location of the ID PROM in the host's address space is dependent on which carrier is used. Normally for VMEbus carriers, the ID PROM space is directly above the IP's I/O space, or at IP-base + \$80.

See the Calibration section below for more information on the stored calibration parameters.

Accessing the ID PROM on ATC4 (AT bus) carrier requires writing to the control register on the carrier board to select the ID PROM space of its IndustryPacks. See the ATC4 manual for more information.

The ID PROM used is an AMD 27LS19A or equivalent.

3F	(available for user)	
31		
2F	DAC C gain_error, low byte	
2D	DAC C gain_error, high byte	
2B	DAC B gain_error, low byte	
29	DAC B gain_error, high byte	
27	DAC A gain_error, low byte	
25	DAC A gain_error, high byte	
23	DAC C offset_error, low byte	
21	DAC C offset_error, high byte	
1F	DAC B offset_error, low byte	
1D	DAC B offset_error, high byte	
1B	DAC A offset_error, low byte	
19	DAC A offset_error, high byte	
17	CRC	
15	No of bytes used	(20)
13	Driver ID, high byte	
11	Driver ID, low byte	
0F	reserved	(00)
0D	Revision	(B1)
0B	Model No IP-16DAC	(25)
09	Manufacturer ID: SBS	(F0)
07	ASCII "C"	(43)
05	ASCII "A"	(41)
03	ASCII "P"	(50)
01	ASCII "I"	(49)

Figure 6 ID PROM Data (hex)

Calibration

This section of the manual describes how to use the calibration data stored in the ID PROM. If an existing driver routine is being used, then the information in this section has already been incorporated, and the user need not worry about calibration further.

For detailed information on accuracy, drift and non-linearity the user is referred to the manufacturer's data sheets for the individual components.

For copy of the factory calibration procedure, which explains in detail how each calibration number is derived, please contact SBS Technologies.

All errors are considered to be linear.

The calibration prom holds all correction numbers in units of 1/4 LSB. The correction numbers are treated as a signed 16-bit word. In this section, all numbers shown are in decimal, unless preceded by a dollar sign (\$). All equations in this section are in units of bits unless otherwise labeled.

When the units are bits the range of the DAC is \$0000 to \$FFFF.

To maintain necessary accuracy in the implementation of these algorithms it is recommended that the programmer use 32-bit signed integers, where 24 bits represents the equivalent ADC reading in bits, and 8 bits represents the fractional LSBs (least significant bits). This permits the use of fast integer arithmetic, without worrying about introduction computational rounding errors, limiting range, or missing overflow/underflow conditions. an alternative is to do all computations in floating point. Note that since the numbers stored in the calibration prom are in units of 1/4 LSB, they require normalization prior to computation.

The numbers in the calibration prom are signed 16-bit integers with units of 1/4 LSB. To normalize to the recommended 32-bit integer data format in the previous paragraph, (i) sign extend from 16-bits to 32-bits, then (ii) shift left arithmetically (keeping the sign) 6 bits.

The Ideal Case

If there were no analog errors whatsoever, then the following would be true. The goal of the correction factors and algorithm is to get as close to this ideal as possible.

1. The on board reference would produce +5.000 and -5.000 volts.
2. Full scale for the DAC is defined as 65,536 counts. For bipolar operation zero is defined as 32,768 (\$8000). Note that since the maximum sixteen bit number is 65,535 (\$FFFF) that the maximum achievable output voltage is one LSB less than the maximum voltage stated in the ideal range. An output of zero to the DAC produces the most negative voltage in the range.

3. For the DAC, one LSB is defined as taking the voltage span of the full scale range and dividing by 65,536.

Range	Span	One LSB
0 to +5 V	5 V	76.2939 μ V
-5 to +5 V	10 V	152.5879 μ V
0 to +10 V	10 V	152.5879 μ V
-10 to +10 V	20 V	305.1758 μ V

DAC

The best way to understand the complete correction formula for the DAC is to follow the process of correcting errors one at a time.

If the gain had no error, but the offset did, then the correction formula would be:

$$(1) \quad \text{Data} = \text{Value} - \text{Offset},$$

where **Data** is the 16 bit number value written to the DAC, **Value** is the desired value (in the range 0 to 65,535), and **Offset** is the correction factor in the PROM, normalized to bits.

However, there is a gain error. The gain correction coefficient is stored as the number of bits (if there were no offset) that the DAC must be corrected at full scale. The formula for unipolar range would be:

$$(2) \quad \text{Data} = \text{Value}(1 + \text{Gain}/65536) - \text{Offset},$$

Where **Gain** is the stored gain error parameter, normalized to bits.

The DAC is calibrated for one range only: -5 to +5V. Bipolar ranges have a zero offset of 32,768. This offset must be subtracted before the gain is corrected, then added prior to outputting the data. This is written:

$$(3) \quad \text{Data} = 32768 + (\text{Value} - 32768)(1 + \text{Gain}/32768) - \text{Offset}.$$

The above formula is complete, although a bit unwieldy, since all numbers must be converted to bits. The formula below is restated, using non-normalized units.

$$(4) \quad \text{Data} = 32768 + (\text{Voltage} * 32768 / 5) * (1 + \text{G_error} / (4 * 32768)) - \text{O_error} / 4.$$

Data is still the 16 bit number that will be sent to the DAC. **Voltage** is the desired output voltage (in the range of -5 to +4.999847). **G_error** and **O_error** are the signed 16-bit non-normalized integers stored in the calibration prom. See Figure 6 above for the locations in the prom. The equation above assumes high (floating point) precision.

Finally, the Data must be range checked for under and overflow conditions. The formula may produce a number less than 0 or more than 65535.

$$(5) \quad \text{Final_Data} = \text{Data} \text{ min } 0 \text{ max } 65535.$$

Final_Data is sent to the DAC.

Other Ranges

Although the ranges other than -5 to +5 are not calibrated, the calibration numbers may be used to improve output accuracy over no calibration at all.

For 0 to +5 use in place of (4):

$$(6) \quad \text{Data} = (\text{Voltage} * 65536 / 5) * \\ (1 + \text{G_error} / (4 * 32768)) - \text{O_error} / 4.$$

For 0 to +10 use in place of (4):

$$(7) \quad \text{Data} = (\text{Voltage} * 65536 / 10) * \\ (1 + \text{G_error} / (4 * 32768)) - \text{O_error} / 4.$$

For -10 to +10 use in place of (4):

$$(8) \quad \text{Data} = 32768 + (\text{Voltage} * 32768 / 10) * \\ (1 + \text{G_error} / (4 * 32768)) - \text{O_error} / 4.$$

Note in all cases that the Data must be range checked (equation 5) before being sent to the DAC.

Uncalibrated Accuracy

The uncalibrated accuracy is computed from the worst case error of the components from their data sheets. It is nominally ±16 bits for the DAC, and ±2 bits for the op amp buffer. The reference voltage is trimmed on the IndustryPack to ±4 bits.

Setting Output Range

Each of the three DACs has four possible output ranges. These are shown in Figure 7 below.

Range	DAC A	DAC B	DAC C	Notes
0 to +5 V	E4.1-E4.2 E5.2-E5.3	E6.1-E6.2 E7.2-E7.3	E8.1-E8.2 E9.2-E9.3	
-5 to +5 V	E4.1-E4.2 E5.1-E5.2	E6.1-E6.2 E7.1-E7.2	E8.1-E8.2 E9.1-E9.2	Factory Default
0 to +10 V	E4.2-E4.3 E5.2-E5.3	E6.2-E6.3 E7.2-E7.3	E8.2-E8.3 E9.2-E9.3	Ext. ±15V supply recommended
-10 to +10 V	E4.2-E4.3 E5.1-E5.2	E6.2-E6.3 E7.1-E7.2	E8.2-E8.3 E9.1-E9.2	Ext. ±15V supply recommended

Figure 7 DAC Output Range Selection

Users are encouraged to review the Calibration section above and the User Options section below before they change DAC output ranges.

As shown in the notes in the Figure above, the factory default option is -5 to +5 V. Calibration data for this range only is stored in the PROM.

The ranges that include -10 or +10 volts benefit from having a higher power supply than the IndustryPack standard ±12 V. If the output drive current is kept low, and the power supplies are at least 12.0 volts, then the 10 volt ranges are usable with the standard supplies.

User Options

The primary user option is settling the output range options for each DAC. See the previous section for this information.

Zero on Reset

If a shunt is installed in position E1, all three DAC channels will be reset to bipolar zero (0V) during hardware reset. If the shunt at E1 is removed, hardware reset will have no effect. If the shunt is removed the power-up state of the DACs is undefined. If the shunt is installed at E1 and one or more DACs is set for a unipolar output range, the DAC will be set to the midpoint of the output range.

External Supplies

External ± 15 volt power supplies are recommended for the two output ranges that include -10 and +10 volts. External supplies may also be used to minimize RF noise and drift on the outputs due to noisy supplies.

Note that both carrier boards and the IP-16DAC have RF filters on the power lines. The AD588 voltage reference is extremely immune to low frequency changes on the input ($\pm 200 \mu\text{V/V}$ worst case). Nonetheless, if significant noise exists at certain frequencies (10 KHz to 1 MHz, typically) on the standard ± 12 volt supplies, better performance will be obtained with dedicated external supplies.

See Figure 8 for shunt positions.

Source	E2	E3	Voltage
Internal	E2-1 to E2-2	E3-1 to E3-2	± 12 V typical from Logic Interface
External	E2-2 to E2-3	E3-2 to E3-3	± 15 V recommended

Figure 8 Power Supply Selection

Remote Sense

Remote sense does not require any shunts to be installed or removed. An internal 100Ω resistor connects the sense input to the analog output at the connector.

Users are encouraged to wire the remote sense line to the analog output line as far “downstream” as possible to effectively prevent cabling resistance from decreasing system accuracy.

Direct Outputs

Users may use the DAC output directly rather than the buffered output. Faster settling time is possible using the direct output.

Remote sense is not available on the direct output. Current drawn from the direct output should be kept very low (in the range of 1 nA) to minimize cabling resistance errors. Note the one LSB is only $75 \mu\text{V}$. Cabling must be kept short to avoid ringing.

Construction and Reliability

IndustryPacks were conceived and engineered for rugged industrial environments. The IP-Precision ADC is constructed out of 0.062 inch thick FR4 material. The six copper layers consist of a shield plane, an output plane, a ground plane, a power plane and two digital signal planes.

The use of a shield plane directly under the ICs as well as full power and ground planes significantly reduces digital noise on the analog outputs. The power and ground planes are split to include both digital and analog portions. The analog and digital grounds are connected at a single point to eliminate ground loops. Three solid copper planes also aid in maintaining constant temperature across the IP. The output traces have their own dedicated layer, with a ground plane on both sides. The output traces are particularly wide to keep output impedance as low as possible. Four connector pins are wired in parallel to reduce contact resistance. Solder mask covers exposed traces on both sides.

Both surface mount and through-hole components are used. Surface mount components are limited to passive components. All ICs are socketed. These sockets are screw-machined pins, gold plated. Note that replacing or swapping positions of *any* IC requires re-calibration. High insertion and removal forces are required, which assists in keeping components in place. If the application requires unusually high reliability or is in an environment subject to high vibration, the user may solder the four corner pins of each IC into the socket, using a grounded soldering iron. Shunts may be replaced with wire-wrap® wires if desired.

After burn-in, units are calibrated with semi-automated procedure using instrumentation traceable to NBS standards. After calibration a custom PROM is made and installed. Final tests are performed, verifying proper calibration and functionality.

The IndustryPack connectors are keyed, shrouded and gold plated on both contacts and receptacles. They are rated at 1 Amp per pin, 200 insertion cycles minimum. These connectors make consistent, correct insertion easy and reliable.

The IP is secured to the carrier with four metric M2 stainless steel screws. The heads of the screws are countersunk into the IP. The four screws provide significant protection against shock, vibration, and incomplete insertion. For most applications they are not required.

The IndustryPack provides a low temperature coefficient of 0.89 W/°C for uniform heat. This is based on the temperature coefficient of the base FR4 material of .31 W/m-°C, and taking into account the thickness and area of the IP. This coefficient means that if 0.89 Watts is applied uniformly on the component side, that the temperature difference between the component and the solder side is one degree Celsius.

IndustryPack Logic Interface Pin Assignment

GND	GND	1	26
CLK	+5V	2	27
Reset*	R/W*	3	28
D0	IDSel*	4	29
D1DMAreq0*	5 30		
D2	MemSel*	6	31
D3n/c	7 32		
D4	n/c	8	33
D5n/c	9 34		
D6	IOSel*	10	35
D7n/c	11 36		
D8	A1	12	37
D9n/c	13 38		
D10	A2	14	39
D11	n/c	15	40
D12	A3	16	41
D13	n/c	17	42
D14	A4	18	43
D15	n/c	19	44
BS0*	A5	20	45
BS1*	Strobe*	21	46
-12V	A6	22	47
+12V	Ack*	23	48
+5V	n/c (+5STBY)	24	49
GND	GND	25	50

Note 1: The no-connect (n/c) signals above are defined by the IndustryPack Logic Interface Specification, but not used by this IP. See the Specification for more information.

Note 2: The layout of the pin numbers in this Figure corresponds to the physical placement of pins on the IP connector. Thus this Figure may be used to easily locate the physical pin corresponding to a desired signal. Pin 1 is marked with a square pad on the IndustryPack.

Interface Pin Assignment

GND Drive	1
n/c	2
DAC A Output (buffered)	3,4,5,6
DAC A Sense input	7
n/c	8
GND Drive	9
n/c	10
DAC B Output (buffered)	11,12,13,14
DAC B Sense input	15
n/c	16
GND Drive	17
n/c	18
DAC C Output (buffered)	19,20,21,22
DAC C Sense input	23
n/c	24
GND Drive	25
n/c	26
DAC A Direct Output	27
n/c	28
GND Drive	29
n/c	30
DAC B Direct Output	31
n/c	32
GND Drive	33
n/c	34
DAC C Direct Output	35
n/c	36
GND Drive	37
V+ External Power (+15 typ)	38,39
GND (for External Power Return)	40,41
V- External Power (-15 typ)	42,43
GND Drive	44,45,46,47,48,49
n/c	50

Specifications

Logic Interface	IndustryPack logic Interface
Number of Channels	Three Digital to Analog Converters
Resolution	16 bits (65,536 codes)
Output Ranges	0 to 5V, 0 to 10V, -5 to +5V, -10 to +10V, pin selectable
Output Options	Direct output (no buffer), Buffer with remote sense
Digital Interface	16-bit write and read-back, I/O and Memory spaces
DAC Clear	Bipolar or unipolar clear on reset or software, pin selectable, may be disabled
Monitonicity	Guaranteed over temperature
Differential Non-linearity	± 1 LSB maximum over temperature
Factory Calibration	Standard, using on board PROM Correction coefficients stored $\pm 1/8$ LSB
Calibrated Accuracy	Zero: ± 2 LSB Gain: $\pm 0.012\%$ of full scale
Drift	± 2 ppm/ $^{\circ}$ C for DAC ± 2 ppm / $^{\circ}$ C for on board reference
Output Impedance	25 m Ω including trace and connector, typical with no carrier or ext wiring
Sense Current	5 nA
Slew Rate	.15 V/ μ sec, typical
Output Settling	70 μ sec typ, large signal, slew rate limited 9 μ sec max to .003% FSR, small signal, DAC settling time limited, 20 μ sec typ, driving 1.0 nF load.
Glitch Energy	400 nV-sec typical over temperature, 100 mV max peak height typical, worst case transition
Wait States	DAC read and write: two, DAC clear: one ID read: zero

Cabling	Four parallel cable lines for each output, One line each for sense inputs, Buffered GND drive between all channels, Direct output lines, GND reference, Four lines external power input
External Power Supply	Pin selectable for external supplies, ± 12 volts from logic interface, standard
Power Requirements	0.9 Watts power dissipation +5 V, 90 mA +12V, 16 mA -12V, 16 mA
Test Conditions	Specifications with output range -5 to +5, Values are typical at 25°C, 12V \pm .25V supply, unless stated otherwise
Dimensions	1.800 by 3.900 by 0.340 inches maximum
Environmental	Operating temperature: 0 to 70°C Humidity: 5 to 95% non-condensing Storage: -10 to +85°C